A DYNAMIC CLAMP PROTOCOL TO ARTIFICIALLY MODIFY CELL

CAPACITANCE

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ABSTRACT

1	Abstract - Dynamics of excitable cells and networks depend on the membrane time constant, set
2	by membrane resistance and capacitance. Whereas pharmacological and genetic manipulations of
3	ionic conductances are routine in electrophysiology, experimental control over capacitance remains
4	a challenge. Here, we present capacitance clamp, an approach that allows to mimic a modified
5	capacitance in biological neurons via an unconventional application of the dynamic clamp technique.
6	We first demonstrate the feasibility to quantitatively modulate capacitance in a mathematical neuron
7	model and then confirm the functionality of capacitance clamp in <i>in vitro</i> experiments in granule cells
8	of rodent dentate gyrus with up to threefold virtual capacitance changes. Clamping of capacitance
9	thus constitutes a novel technique to probe and decipher mechanisms of neuronal signaling in ways
10	that were so far inaccessible to experimental electrophysiology.

11 **1 Introduction**

Membrane capacitance is a major biophysical parameter in neurons and other excitable cells, which determines how 12 fast the membrane potential changes in response to a current [1, 2]. How capacitance impacts electrical signaling and 13 neuronal processing, however, can rarely be observed experimentally, because its value appears to be constant for most 14 membranes, around 1.0 uF/cm2 [3]. A prominent exception is the reduced capacitance of myelinated axons, which 15 allows faster action potential propagation [4] and thereby crucially contributes to cognitive functions [5]. The effects 16 of capacitance changes can, therefore, so far only be compared via mathematical simulations, where capacitance is 17 simple to control. Such modeling, for example, suggests that the reduced membrane capacitance observed in human 18 pyramidal cells can serve to increase synaptic efficacy [6; but see 7]. Nevertheless, experimental manipulation of 19 capacitance remains challenging; in particular because changes in membrane area, thickness and lipid composition that 20 affect capacitance might influence other membrane functions, such as the embedding of ion channels, with potentially 21 unintended and uncontrolled consequences for electrical behavior. Here, we address this technical challenge by 22 introducing capacitance clamp (CapClamp): an intracellular recording mode based on the dynamic clamp that emulates 23 altered capacitance values in biological neurons [8, 9]. Via CapClamp, the voltage dynamics governed by the actual 24 biophysics of a cell – active ion channels and synaptic inputs – can thus be flexibly probed under multiple "virtual" 25 capacitance conditions, which provides precise experimental control over this hitherto inaccessible parameter. 26

In addition to the analysis of biological capacitance adaptations, control over capacitance offers a distinct way to probe 27 cellular electrical dynamics. Capacitance has a unique temporal role, because its direct effects are restricted to the 28 membrane time constant whilst leaving the steady state I-V curve unaltered. In this way, capacitance differs from leak 29 conductance, the other determinant of the time constant, which also alters steady-state response amplitudes. For this 30 reason, theoretical studies preferentially vary capacitance to investigate ion channel dynamics [10, 11] and qualitative 31 switches (bifurcations) in neural excitability [12, 13]. Furthermore, effects of an altered capacitance can be informative 32 about more complex, time scale-related parameters like temperature or ion concentrations [14]. Such computational 33 predictions, however, often rely on simplified neuron models, so a similar experimental control over capacitance would 34 be desirable to test them in biological cells. 35

The proposed CapClamp alters capacitance in a virtual manner, combining the simplicity of computational control with 36 the complex biophysics of a real neuron. It is inspired by the dynamic clamp technique, which has originally been 37 developed to simulate the presence of additional conductances in a biological neuron relying on a fast feedback loop 38 between intracellular recording and a computational model [8, 9, 15, 16]. The precise control over virtual conductances 39 enables electrophysiological experiments that are more difficult or even impossible with traditional pharmacological 40 or genetic means [17-22]. Here, we demonstrate how the dynamic clamp can be extended to virtual capacitance 41 modifications by currents designed to speed up or slow down dynamics of the membrane potential. We derive a simple 42 expression for these CapClamp currents, which can be applied in all excitable cells and only requires the experimenter 43 to specify the original cell and the desired target capacitance. In an experiment based on a hardware-implemented 44

RC circuit, we verify that the CapClamp indeed correctly modifies the time constant. Via numerical simulations, we confirm that a clamped model neuron exhibits the same pronounced changes of firing and spike shape as a control cell with an altered capacitance. For an experimental demonstration, we clamp the near-somatic capacitance of rat dentate gyrus granule cells and analyze how the induced local capacitance change affects their spiking behavior. Finally, we illustrate how the CapClamp can be used to probe signal integration, energy consumption and bifurcations of excitable cells in ways that so far were experimentally inaccessible.

51 2 Results

52 2.1 Capacitance clamp: a dynamic clamp protocol to mimic capacitance changes

Dynamic clamp relies on a fast feedback loop between an intracellular recording of a neuron and a computer that 53 simulates virtual cellular or circuit components online. Originally, the dynamic clamp has been developed to study 54 how a membrane conductance alters the neuron's voltage dynamics [8, 9]. In each recording interval (i.e. time interval 55 between two voltage samplings), a digital model of the conductance receives the sampled membrane potential, updates 56 the conductance state and sends the corresponding current value back to the amplifier. In this way, given a sufficiently 57 high update rate $f_{\rm dyn}$ (often ≥ 10 kHz), the current through the recording electrode accurately mimics the current 58 through the modeled conductance and the dynamics of the neuron appear as if this conductance was physically present 59 in the membrane. 60

⁶¹ Whereas conductances gate ionic currents across the membrane, the capacitance determines how fast these currents ⁶² can change the membrane potential. Every altered membrane property that results in a modified capacitance value, ⁶³ such as membrane area, thickness or lipid composition, affects this rate of change of the membrane potential (Fig. 1 A). ⁶⁴ To artificially mimic a modified capacitance, we therefore first asked whether a dynamic clamp protocol with its fast ⁶⁵ feedback loop between voltage sampling and current injection could adjust the "speed" of a cell's membrane potential ⁶⁶ (Fig. 1 B). Using the current balance equation, the basic mathematical description of membrane voltage dynamics, we ⁶⁷ derived a capacitance clamp (CapClamp) scheme with a simple expression for the clamping current I_{dyn} (see Methods),

$$I_{dyn,i} = \frac{C_c - C_t}{C_t} \left(C_c \frac{V_i - V_{i-1}}{\Delta t} - I_{dyn,i-1} \right),$$
(1)

which only requires the experimenter to measure the cell capacitance C_c in order to set a new target capacitance C_t . In every recording interval $\Delta t = f_{dyn}^{-1}$, the CapClamp uses the measured cell capacitance value C_c and the voltage derivative $\frac{V_i - V_{i-1}}{\Delta t}$ to estimate the present membrane current and then increases ($C_t < C_c$) or decreases ($C_c < C_t$) the net current by insertion of a correction current in the next time bin. In this way, despite a physically unaltered capacitance, the membrane potential changes faster or, respectively, slower – as if the clamped cell actually had the different capacitance C_t selected by the experimenter. In the following, we will demonstrate the CapClamp in simulated

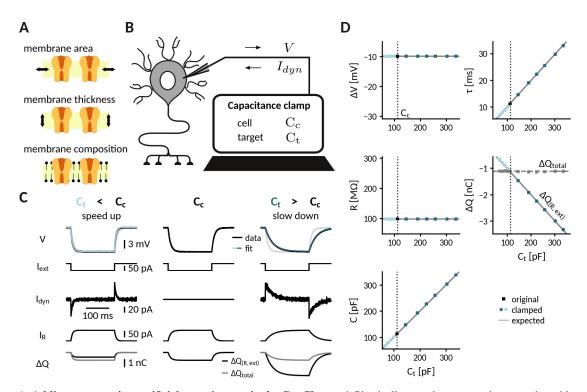


Figure 1: Adding or removing artificial capacitance via the CapClamp. A Physically, membrane capacitance varies with surface area, thickness and lipid composition **B** Virtual capacitance modification via the CapClamp is a form of dynamic clamp, a fast feedback between intracellular voltage sampling and computer-controlled current injection: given the measured cell capacitance C_c , the target capacitance C_t and recorded membrane potentials, the computer calculates clamping currents required to mimic the desired change of capacitance. **C** Clamping a hardware-implemented model cell (RC circuit) at a decreased (left) or increased (right) capacitance leads to faster respectively slower charging of the "membrane potential" V (top row, exp: black traces, exponential fit:light and dark blue) in response to a step current I_{ext} (2nd row) due to the clamping currents I_{dyn} (3rd row). As a result, the current through the resistance $I_R = -\frac{V}{R}$ (4th row) has a different profile and the apparently deposited charge $\Delta Q(R, ext) = \int dt I_R + I_{ext}$ (5th row, black) by the "cellular" transmembrane currents decreases, respectively, increases as expected for a capacitance change. The total deposited charge ΔQ_{total} (gray), taking into account the clamping currents, is the same in all three cases, because the physical capacitance did not change. **D** Measured time constant τ , voltage responses ΔV , resistance R, deposited charge ΔQ (apparent and total) and capacitance C versus target capacitances.

- ⁷⁴ and experimental scenarios with increasing complexity ranging from a passive RC circuit up to biological neurons with
- ⁷⁵ a spatially extended morphology.

76 2.2 Clamping capacitance in a passive cell

- ⁷⁷ The simplest scenario to apply the CapClamp is a single compartment passive cell, equivalent to an RC circuit. In the
- $_{78}$ absence of active conductances, the effects of a capacitance change can be precisely formulated: the capacitance C sets
- ⁷⁹ the membrane time constant $\tau = RC$, determining how fast the membrane potential changes in response to a current.
- $_{80}$ Note that, in contrast to the resistance R, the change in capacitance leaves the voltage amplitude of the steady-state
- ⁸¹ response unaltered. To quantitatively confirm the effects of clamping capacitance and the ability of an exclusively
- temporal control, we measured time constant and capacitance of a clamped RC circuit in experiment and analyzed the
- temporal filtering properties of a modeled clamped circuit using mathematical analysis.

To experimentally characterize a clamped passive cell, we implemented the CapClamp scheme in a dynamic clamp setup 84 (see Methods) and recorded voltage responses to current pulses from the simplest possible model cell, i.e., a hardware 85 implemented RC circuit, while clamping it at a range of target capacitances (Fig. 1 C). As expected for an RC circuit, 86 the charging curve of the unclamped model cell was fit well by a single exponential, whose time constant (τ =11.1 ms) 87 and voltage amplitude (ΔV =-9.9 mV) allowed us to determine the circuit's resistance R=99.4 M Ω and capacitance 88 C=112.3 pF. This capacitance value was then used as the cell capacitance C_c input for the CapClamp. Clamped 89 at a decreased capacitance, the time constant shortened (C_t =67.4 pF: τ =6.6 ms) and at an increased capacitance, 90 it lengthened (C_t =336.9 pF: τ =33.0 ms), but in both cases the steady state voltage amplitude remained the same. 91 Accordingly, the measured capacitance of the clamped circuit confirmed the chosen target capacitance for the whole 92 tested range from a 0.6- up to a 3-fold change with respect to the original capacitance (e.g. C_t =67.4 pF: C=67.5 pF; 93

⁹⁴ C_t =336.9 pF: C=338.1 pF), whereas the measured resistance remained constant (Fig. 1 D).

As a consequence of the correctly transformed voltage response, the leak current in the clamped RC circuit also behaved 95 as if the capacitance had changed. When the circuit was clamped, the leak current through the resistance, $I_R = \frac{V}{R}$, 96 exhibited a shorter ($C_t < C_c$) or longer ($C_t > C_c$) transient until reaching steady state. As a consequence, the charge 97 $\Delta Q(I_R, I_{ext})$ deposited on the capacitance by the apparent "transmembrane" current, the sum of leak and external 98 stimulus current, reduced ($C_t < C_c$) or increased ($C_t > C_c$) to the extent expected for an altered capacitance (Fig. 1 99 C). In contrast, the overall deposited charge $\Delta Q(I_R, I_{ext}, I_{dvn})$, including the clamping current, was identical in the 100 clamped and the original circuit, reflecting that the physical capacitance did not change. For the simple RC circuit 101 considered here, the distinction between the clamping current and the intrinsic "cellular" currents might appear artificial, 102 because all currents use the same charge carrier. In a biological neuron, however, this distinction becomes relevant, 103 because the clamping currents through the recording electrode might rely on other charge carriers (depending on the 104 used intracellular solution) than the cellular currents governed by multiple ion selective channel types. 105

For more complex stimuli than a simple current pulse, the temporal filtering properties of a clamped membrane 106 determine how well the CapClamp mimics the chosen capacitance change. To generally assess these filtering properties, 107 we analytically derived the frequency-dependent impedance of a modeled clamped RC circuit using linear control 108 theory (Fig. 1 - suppl. 1 A, see Appendix). The derived impedance profiles confirmed the experimentally observed 109 altered time constants. For example, an RC circuit clamped at an increased capacitance further attenuated non-zero 110 frequencies reflecting its longer time constant. Overall, impedance amplitudes of a clamped RC and the corresponding 111 target circuit fit well up to a tenth of the dynamic clamp frequency f_{dyn} , that is up to ≈ 2 kHz for a 20 kHz dynamic 112 clamp system as used here (Fig. 1 - suppl. 1 B and C). As high frequencies are heavily attenuated by the low pass filter 113 of a cell's membrane, these differences lead to relatively small deviations in the voltage responses. The mathematical 114 analysis thus suggests that for a fast dynamic clamp system (>20 kHz), the CapClamp is expected to work well for most 115 stimuli with time scales in the physiological range. 116

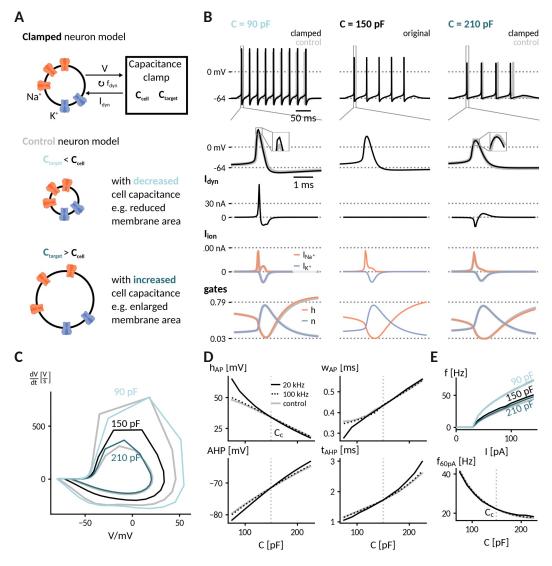


Figure 2: Simulation of the capacitance clamp in a conductance based neuron model. A Neurons coupled to the CapClamp are compared with control neurons with an altered capacitance (depicted as a difference in membrane area). B Spiking at 0.6-fold decreased (90 pF), original (150 pF) and 1.4-fold increased capacitance (210 pF) with from top to bottom: spike shape, dynamic clamp current, ionic currents (Na⁺, K⁺) and gating states (h: sodium inactivation gate, n: potassium activation gate). Clamped and original traces in black or color, control in gray. C Comparison of spike shapes in the V- $\frac{dV}{dt}$ -plane (black: original, blues: clamped, gray: control). D Comparison of spike amplitude h_{AP} (top left), spike width w_{AP} (top right), after hyperpolarization amplitude AHP (lower left) and timing t_{AHP} (lower right) across different capacitances in B. *Bottom*: Comparison of firing frequency at 60 pA across different capacitances (solid: 20 kHz, dotted: 100 kHz, gray:control).

117 2.3 Simulation of the CapClamp in a biophysical neuron model

In neurons with active spike-generating conductances, capacitance changes impact neuronal firing via the interplay 118 of the altered membrane time constant and the gating kinetics of the channels involved. As gating dynamics can be 119 in the sub-millisecond range, e.g. for transient sodium channels, the CapClamp is expected to require a sufficiently 120 high dynamic clamp frequency to accurately reproduce changes of spike shape or firing rate. To understand these 121 requirements and lay the ground for investigations of capacitance changes in biological neurons, we simulated the 122 CapClamp in a neuron model with biophysical channel dynamics and a single-compartment morphology (see Methods). 123 The simulation allowed us to compare the firing of the clamped neuron to the expected firing at this modified capacitance. 124 Specifically, we inspected the spiking responses to a depolarizing current for the original 150 pF, a decreased 90 pF and 125 an increased 210 pF capacitance, for the latter two comparing clamped and expected dynamics (Fig. 2 A). Capacitance 126 changes exerted a notable influence on both firing frequency and spike shape, which was mostly well-captured by 127 the simulated CapClamp (Tab. 1). When the capacitance was decreased to 90 pF, spiking frequency speeded up and 128 action potentials had a larger peak amplitude, a decreased duration and an increased afterhyperpolarization (AHP). 129 When the capacitance was increased to 210 pF, the effects were opposite: spiking frequency slowed down and action 130 potentials had a smaller peak, an increased duration and a reduced AHP. At increased capacitances, spike amplitudes of 131 the clamped neuron were larger than expected, a consequence of the limited tracking of the fast sodium current at the 132 used dynamic clamp frequency (Fig. 2 B). Except for this brief overshoot, the CapClamp overall forced the membrane 133 potential on the expected trajectory and correctly adjusted the resulting ionic currents and the gating variable dynamics 134 of the active conductances, e.g. for a reduced capacitance of 90 pF, the sodium current became narrower in time and 135 exhibited a second peak (Fig. 2 A). 136

A subsequent comparison of simulated spiking for the whole range of tested target capacitances from 75 pF to 225 pF confirmed that the CapClamp reliably reproduced the main effects of a modified capacitance on spike shape (Fig. 2 C) and firing frequency (Fig. 2 D). As expected, a crucial factor for a good quantitative fit is the dynamic clamp frequency – observable differences at a 20 kHz sampling frequency were strongly reduced for a sampling frequency of 100 kHz (Fig. 2 C and D). In this regard, the chosen neuron model is especially demanding because its rapid gating dynamics are fit to a fast spiking interneuron [23]. Taken together, our simulations show that capacitance impacts neuronal spiking from firing frequency to action potential shape and that the CapClamp is well-suited to study these effects.

Table 1: Firing frequency and spike shape in a biophysical neuron model for a decreased, the original and an increased capacitance, comparing simulations of an actually altered capacitance with the CapClamp. Values are shown as actual(**clamped**).

C (pF)	f (Hz)	$\mathrm{h_{AP}}\left(mV\right)$	w_{AP} (ms)	AHP (mV)
decreased 90	34.9(34.3)	45.7(55.0)	0.30(0.30)	-77.8(-79.7)
original 150	22.1	33.9	0.39	-71.5
increased 210	17.8(18.9)	21.4 (20.1)	0.48(0.48)	-66.0(-64.7)

144 **2.4** Experimental demonstration of the CapClamp in rat dentate gyrus granule cells

Biological neurons differ from the simple "cells" considered so far, i.e. RC circuit and single compartment neuron 145 model, in a major aspect: they can have complex morphologies, where the membrane potential varies between different 146 compartments and membrane capacitance is distributed across the neuronal structure. As the CapClamp in contrast 147 operates locally through the recording electrode, the emulated capacitance change is expected to be localized to the 148 recorded compartment instead of affecting all compartments (see Methods). To demonstrate such localized capacitance 149 changes and study their effects on neuronal spiking, we applied the CapClamp in *in vitro* patch-clamp recordings of rat 150 dentate gyrus granule cells (DGGCs). Among morphologically complex cells, DGGCs appear well-suited to test the 151 CapClamp, because their morphological structure, consisting of a central soma and one to four primary apical dendrites 152 as shown in Figure 3 A [24], translates to a relatively compact electrotonic structure [25, 26]. 153

154 2.4.1 Measurement of local near-somatic capacitance

Most capacitance measurements aim to provide an accurate estimate of the *global* capacitance of a neuron [1, 2]. To 155 correctly infer the transmembrane and axial current, however, the CapClamp requires the local capacitance value of the 156 compartment where the electrode is placed at. For the somatic DGGC recordings, we exploit that the current clamp step 157 method – fitting charging curves via a sum of exponential terms – can also provide local capacitance information [1]. 158 DGGC charging curves consisted of a slow (τ_0 : 15.1 ± 4.8 ms, R_0 : 127 ± 45 MΩ) and a fast (τ_1 : 0.77 ± 0.24 ms, R_1 : 159 $35 \pm 15 \,\mathrm{M\Omega}$) component. Such a response can be understood in terms of a two compartment circuit consisting of a 160 *near* compartment, comprising the patched soma and its surrounding, coupled to a *far*, mostly dendritic, compartment 161 as depicted in Figure 3 A (for details on the mapping, see Methods). Importantly, the slow and fast components 162 can be mapped to the corresponding five circuit parameters: near capacitance C_n (21.0 ± 9.4 pF), near resistance 163 R_n (854 ± 394 MΩ), coupling resistance R_a (53 ± 20 MΩ), far capacitance C_f (106 ± 33 pF) and far resistance R_f 164 $(156 \pm 60 \text{ M}\Omega)$ (Fig. 3 C). Accordingly, this near-somatic capacitance C_n represents the summed capacitance of the 165 membrane area that is electrotonically close to the recording site and thus is the value that the CapClamp requires as 166 input and should be able to modify. 167

168 2.4.2 Altered near-somatic capacitance in DGGCs

To confirm the localized effect of the CapClamp, we repeated the above capacitance measurement while clamping DGGCs at values ranging from 0.6 to 3 times the original near capacitance. Figure 3 B depicts how the charging of the membrane potential in an exemplary cell changed its shape in reaction to the clamp. Both slow and fast time constant lengthened with capacitances, whereas the associated resistances increased and decreased, respectively, such that their sum, the total input resistance (which is expected to be independent of capacitance), remained constant. These measured time constants and amplitudes matched the predicted ones for a two compartment circuit with a near capacitance at the chosen target values and all other circuit parameters at their original values. In a multicompartment simulation of a

morphologically reconstructed DGGC, we could reproduce both the two compartment structure of DGGCs and the
 isolated modification of the near capacitance, further confirming the local control via the CapClamp.

Across 18 recorded cells, the CapClamp robustly altered DGGC charging curves and allowed to modify the charging 178 time constants. Within the tested capacitance range, the slow time constant τ_0 decreased by -0.8 (-1.0 to -0.6) ms, 179 median and interquartile range in parentheses, and increased up to 3.0 (2.4 to 3.9) ms, whereas the fast time constant 180 τ_1 changes ranged from -0.24 (-0.29 to -0.20) ms up to 0.60 (0.36 to 0.86) ms (Fig. 3 D). To quantify how well these 181 changes reflected an altered near capacitance, we evaluated the goodness of fit between the observed and expected time 182 constants and resistances. In the majority of cells, R-squared values were close to 1, indicating that the CapClamp 183 induced the expected changes (τ_0 : 0.87 (0.76 to 0.92), R_0 : 0.77 (0.56 to 0.89), τ_1 : 0.76 (0.32 to 0.97), R_1 : 0.85 184 (0.75 to 0.91)). The largest mismatches occurred for the fast time constant, especially at high capacitances, where the 185 measured time constant was often shorter than predicted (Fig. 3 D). A small bias towards a shorter fast component is 186 to be expected and also present in the multicompartment simulation, because this time constant was only about ten 187 times longer than the recording interval of 50 us limiting its slow-down by the CapClamp currents. Larger deviations of 188 τ_1 however could not be reproduced in numerical simulations and likely result from other error sources, such as the 189 difficulty of fitting this small and short time constant in the presence of noise or imprecise estimates of the original near 190 capacitance (see Methods). Overall, in terms of circuit parameters, the capacitance measurements overall confirmed 191 the targeted near capacitance change for 12 out of 18 cells within an average error of 10% (Fig. 3 E). Concluding, the 192 CapClamp achieved an isolated change of the near-somatic capacitance in DGGCs and thereby allows to control the 193 time constants of their passive voltage dynamics. 194

195 2.4.3 Near-somatic capacitance governs action potential shape and firing frequency in dentate gyrus granule 196 cells

In neurons such as the recorded DGGCs, where the axon directly emerges from the soma, the ability to clamp the near-somatic capacitance provides control over the major capacitive load for the action potential generating site in the axon initial segment. Consequently, the CapClamp, although acting locally, is expected to impact action potential (AP) dynamics and excitability of a morphologically complex DGGC as demonstrated earlier for the simplified singlecompartment neuron model (Fig. 2). To illustrate how the CapClamp can thus be applied as as a novel probe to characterize neuronal firing, we compared spiking responses and f-I curves across near capacitances ranging from 0.6 to 3 times the original value, corresponding to a range from 10 pF to 60 pF.

Clamping the near-somatic capacitance in DGGCs, we observed pronounced changes in the spiking response to depolarizing current step, clearly visible in the raw voltage traces (Fig. 4 A). The most apparent change was an altered AP shape (Fig. 4 B) – a continuous reduction of AP peak amplitude (from 61 ± 9 mV at $0.6 C_n$ to 24 ± 16 mV at 3 C_n) and a simultaneous broadening of AP width (from 0.75 ± 0.15 ms at $0.6 C_n$ to 1.30 ± 0.40 ms at $3 C_n$) with increasing capacitance (Fig. 4 C and D). In addition, fast afterhyperpolarization (fAHP) was diminished and disappeared in the majority of cells after increasing capacitance (fAHP in 9/10 cells at $0.6 C_n$ and 2/10 at $3 C_n$). Importantly, the

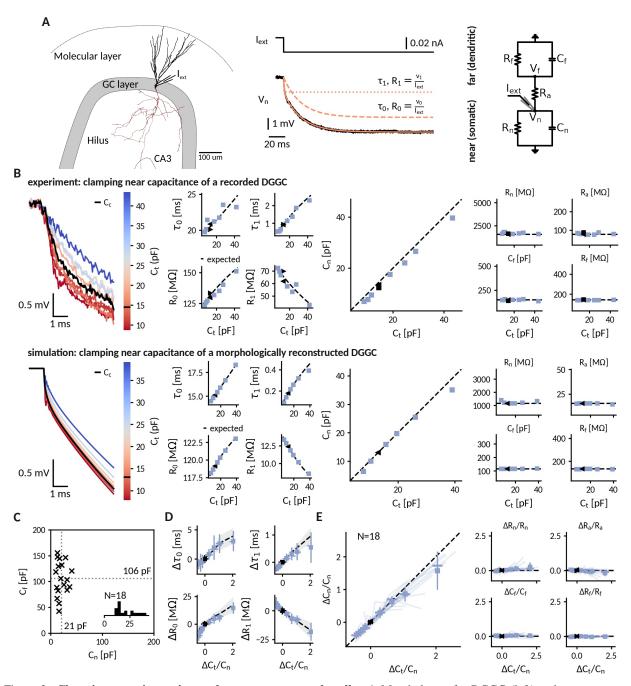


Figure 3: Clamping capacitance in rat dentate gyrus granule cells. A Morphology of a DGGC (left) and response to a hyperpolarizing current injected at the soma, fit via a sum of exponential terms with a slow τ_0 , v_0 and a fast component τ_1 , v_1 (middle), which can be mapped to two resistively coupled RC-circuits (right) with a near (somatic) compartment C_n and R_n , resistive coupling R_a and a far (dendritic) compartment C_f and R_f . B Left: Voltage responses of a recorded (top) and a simulated morphologically-reconstructed (bottom) DGGC to a current pulse (exp: -27 pA, sim: -50 pA) clamped at 0.6- to 3-fold the cell's near capacitance (black: original, color: target near capacitances). Middle: Slow and fast components versus target capacitance. \blacktriangleleft , \triangleright : before and after clamping, blue square: clamped, dashed line: expected values. C Measured near C_n and far C_f capacitances for 18 DGGCs (gray dotted: mean). Inset: histogram of near capacitances. D Changes of slow and fast components in all recorded cells versus relative targeted change of near capacitance (squares: rean, horizontal line: median, vertical line: std, shaded area: std of expected changes). E Relative changes of circuit parameters versus relative.

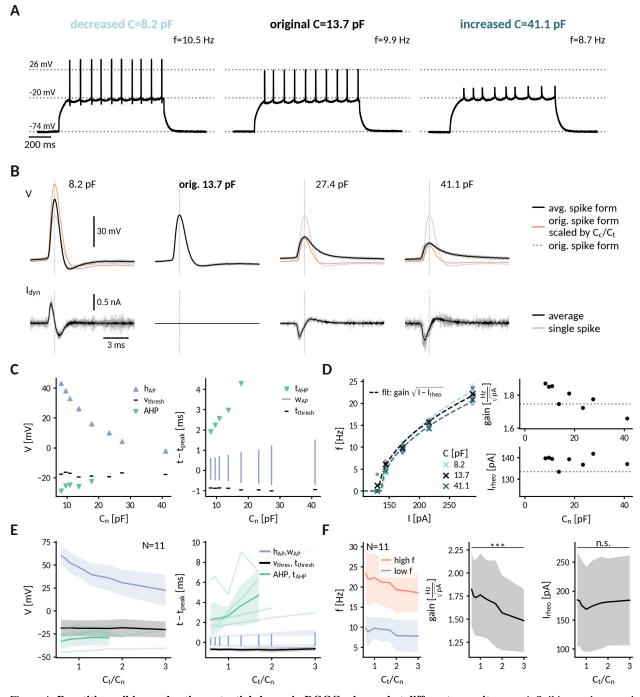


Figure 4: **Repetitive spiking and action potential shapes in DGGCs clamped at different capacitances.** A Spiking at decreased 0.6-fold (left), original (middle) and increased 3-fold (right) near capacitance C_n . **B** Spike shapes (top) and capacitance clamp currents (bottom) for increasing capacitances from 0.6 to 3-fold of the original near capacitance (black: mean, light gray: single spikes, orange: expected spike shape for unaltered intrinsic currents, dotted: spike shape at original capacitance). **C** Comparison of spike shape (left) and temporal structure (right) across tested near capacitances. **D** Measured f-I at 0.6, 1 and 3-fold near capacitance with fit $f = gain\sqrt{I - I_{rheo}}$ (dashed lines). Extracted gain and rheobase for all tested near capacitances (dotted line: values at original capacitance 13.7 pF). **E** Effect of capacitance changes on spike shape (left) and temporal structure (right) for all recorded DGGCs (solid: mean, shaded: std). To compare different cells, the capacitance is shown relative to the original near capacitance and spikes were compared at 1.2 fold of the cell's rheobase. **F** Effect of capacitance changes on firing frequency, low firing (blue) at 1.2 fold rheobase (left), gain (middle) and rheobase (right) for all recorded DGGCs (solid: mean, shaded: std).

observed disappearance of fAHP cannot be explained by increased capacitive filtering alone, which would decrease the its amplitude, but cannot abolish it. It thus demonstrates that the somatic capacitative load in DGGCs is able to influence the AP generating currents.

To illustrate the interplay of capacitance and the AP generating currents, we compared the observed spikes with 213 hypothetical ones obtained by assuming unaltered currents with respect to the original capacitance (see Methods). 214 Recorded and hypothetical spike shapes exhibited marked differences (Fig. 4 B). At 0.6-fold decreased capacitances, 215 for example, the recorded AP amplitude was significantly smaller than the hypothetical one (rec.: 61 ± 9 mV, hyp.: 95 216 \pm 18 mV, Wilcoxon signed-rank p<0.001), presumably reflecting a reduction of the driving force for the sodium current 217 when the AP peak approaches the reversal potential of sodium. Furthermore, at 3-fold increased capacitance, as noted 218 above the recorded spikes exhibited no fAHP in most cells while the hypothetical ones still did (fAHP rec: 2/10, hyp: 219 10/10) – potentially a result of a reduced activation of potassium channels due to lower AP amplitudes and/or earlier 220 closing during the slowed AP repolarization. In contrast to driving force and gating dynamics, the channel kinetics, 221 e.g. their activation curves, cannot be altered by capacitance. Correspondingly, the spike threshold, which reflects the 222 voltage where sodium channels start to massively open, was not significantly correlated with near capacitance (Pearson 223 correlation r=0.089, p=0.42). Taken together, our analysis indicates that an altered somatic capacitance affects both 224 sodium and potassium currents in DGGCs. 225

Controlling the spike initiation zone, near-somatic capacitance also governed DGGC excitability. With increasing 226 capacitance, DGGCs became less excitable and firing frequencies significantly decreased (Fig. 4 D and F). From 0.6- to 227 3-fold of the original near capacitance, the decrease was modest for low firing rates close to threshold (from 9.2 ± 3.6 228 Hz to 7.6 \pm 3.7 Hz, Wilcoxon signed-rank Z=47, p=0.024) and became more pronounced for high firing rates at the 229 largest injected currents (from 22.2 ± 7.0 Hz to 17.9 ± 5.1 Hz, Z=55, p=0.001). In this respect, the firing rate-current 230 (fI) curves of the DGGCs resembled those obtained for the simulated neuron (compare Fig. 2 D): the gain decreased for 231 increased capacitances (from 1.74 \pm 0.45 Hz/ $\sqrt{\text{pA}}$ at 0.6 C_n to 1.42 \pm 0.37 Hz/ $\sqrt{\text{pA}}$ at 3 C_n , Z=55, p=0.001), but the 232 rheobase current remained relatively constant (from 182 ± 77 pA at 0.6 C_n to 179 ± 75 pA at 3 C_n , Z=12, p=0.13). 233 Overall, we conclude that a change of the near-somatic capacitance alone was sufficient to modify the input-output 234 relationship of the recorded DGGCs. 235

236 2.5 Applications of the CapClamp

The CapClamp lends itself to either test hypotheses on the impact of capacitance or to exploit the control over the membrane time constant in order to to alter neuronal dynamics in informative ways. In the following, we briefly illustrate use cases of the CapClamp from these two fields, applying the technique to experimentally explore effects of capacitance on temporal integration, energetic costs of spiking and bifurcations.

241 2.5.1 Temporal integration

A basic processing step in neuronal computation is temporal integration, the summation of time-separated synaptic 242 inputs [27, 28]. An upper limit for temporal integration, at least in the absence of dedicated active channels, is set by 243 the membrane time constant $\tau = RC$, which is directly proportional to the cell's capacitance. Hence, increasing the 244 capacitance of a cell should make it a better integrator: if two inputs arrive separated in time, the cell's response to 245 the second one should be higher than to the first one. Indeed, when we compared the responses of DGGCs to current 246 pulse trains with different interstimulus intervals (ISI) clamped at decreased and increased near-somatic capacitances, a 247 capacitance increase allowed the cell to better "sum" pulses at an ISI of 5 ms as apparent by the larger step sizes in 248 the stair-like voltage response and the finally higher ratio of last to first pulse response. (Fig. 5 B). At an ISI of 50 249 ms, in contrast, neither capacitance allowed temporal integration. The biological relevance of tailoring capacitance 250 to temporal processing can, for example, be observed in auditory cells of the barn-owl, which have no dendrites to 251 reduce capacitive and resistive load and hence shorten their time constant such that they can perform sub-millisecond 252 coincidence detection [29]. 253

254 2.5.2 Energy consumption during spiking

Action potentials are energetically expensive, because the required sodium and potassium ions have to be pumped back 255 using ATP [20, 30]. The minimal amount of ionic charge required for an action potential is dictated by the capacitance 256 as $Q = C\Delta V_{AP}$, suggesting that a smaller capacitance is energetically favorable. In order to gauge how capacitance 257 affects charge deposit and energy consumption, we reexamined spike shapes for a fixed current input at different 258 capacitances both in the simulated neuron and in the recorded DGGCs (Fig. 5 C and E). We found that despite a reduced 259 amplitude at higher capacitances, these smaller spikes still required more depolarizating charge $Q_+ = C\Delta V_{AP}$ (Fig. 5 260 D and F). In the model, we tested whether this depolarizing charge provided a reliable indication of the sodium charge 261 $Q_{\rm Na^+}$, which finally determines pump activity and energy consumption (Fig. 5 D). The sodium charge exceeded the 262 depolarizing charge, because sodium and potassium currents temporally overlap, but it increased in a similar manner 263 with capacitance. Taken together, in the tested model and the recorded DGGCs, energy consumption per action potential 264 appears to be reduced at smaller capacitances. In line with this observation, it has been reported that perineuronal nets 265 could decrease membrane capacitance of fast-spiking interneurons, thereby facilitating high-frequency firing, while 266 keeping energetic costs at bay [31]. 267

268 2.5.3 Neuronal bifurcations

To optimally support neural processing, nerve cells exhibit qualitatively different response properties, which in some cases can be flexibly adapted to context. For example, neurons with class 2 firing (non-zero minimum frequency) can be switched to class 1 firing (arbitrarily low frequency) via neuromodulation [32, 33], transforming them from resonators to integrators. These qualitative differences in response and processing properties are linked to distinct spike generation mechanisms, whose corresponding excitability classes – as well as transitions between them – can be characterized

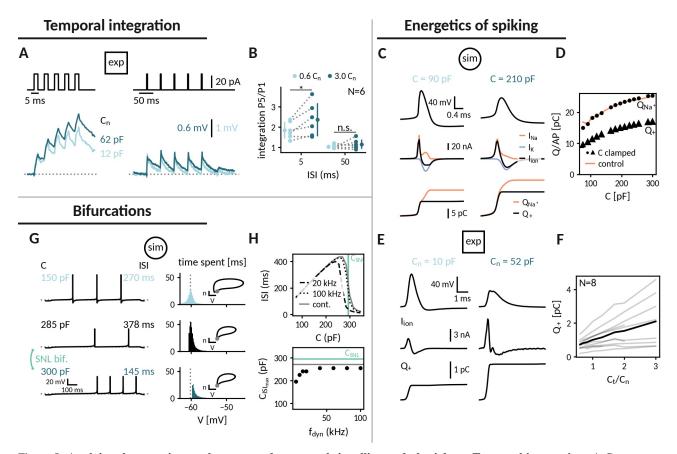


Figure 5: Applying the capacitance clamp to study neuronal signalling and physiology. Temporal integration: A Current pulses with interstimulus intervals of 5 ms and 50 ms (top) and voltage responses of an exemplary DGGC at a decreased (12 pF) and an increased (62 pF) near capacitance (voltage scale adapted to first response height). B Ratio of fifth and first response as a measure of temporal integration for a 0.6-fold decreased capacitance in comparison to a 3-fold decreased one at 5 ms and 50 ms ISI. Energetics of spiking: C Spike shape (top), sodium, potassium and total ionic current (middle) and deposited sodium Q_{Na_+} as well as depolarizing Q_+ charge (bottom) in the Wang-Buzsaki neuron model for a 90 pF and a 210 pF capacitance. D Sodium Q_{Na_+} and depolarizing Q_+ charge per action potential versus capacitance (original: 150 pF, clamped: dot and triangle, control: gray). E Spike shape and depolarizing charge for a dentate gyrus granule cell clamped at decreased 10 pF and increased 52 pF capacitance. F Deposited depolarizing charge versus relative change of near capacitance in recorded DGGCs (black: mean, gray: individual cells). Studying bifurcations: G Left: In the Wang-Buzsaki neuron model, spiking slows down when increasing capacitance up to 285 pF, but a further small increase to 300 pF abruptly doubles the firing frequency, a signature of a saddle-node-loop (SNL) bifurcation. *Right*: Histograms of the time spent at different membrane potentials between two spikes. Insets show the spike dynamics in the potassium gating n-voltage V plane. H Capacitance clamp allows to induce the SNL bifurcation and locate the critical capacitance C_{SNL} . *Top*: Interspike interval against capacitance for simulated capacitance clamps at 20 kHz (dashed) and 100 kHz (dotted) dynamic clamp frequency versus the control simulation (gray) with indication of the expected critical capacitance C_{SNL} (green). *Bottom*: Capacitance at maximal ISI against dynamic clamp frequency.

by bifurcation analysis [34, 35]. Prior work recently highlighted the *saddle-node loop* (SNL) bifurcation, because it separates firing regimes with drastic differences in synchronization and can be induced by a wide array of physiological parameters, including temperature and extracellular potassium, as well as capacitance [13, 14]. We chose this transition and asked whether the CapClamp can correctly locate the SNL bifurcation and, therefore, provide an estimate where a neuron "positions itself" with respect to this critical switch.
To detect the SNL bifurcation in capacitance space, we looked for one of its footprints: frequency-doubling, a drastic

increase of frequency for an incremental capacitance increase. In the simulated neuron, testing capacitances from 150 pF 280 up to 285 pF, the neuron fires at ever slower rates, but at 300 pF, the neuron indeed abruptly doubles its firing rate (Fig. 5 281 G). At this capacitance, the neuron fires faster, because the modified action potential switches to a trajectory, where 282 it skips one half of the slow recharging after a spike (see the voltage histograms in Fig. 5 G). In the simulations, the 283 CapClamp reproduced this frequency-doubling and, given a sufficiently fast dynamic clamp (>40 kHz), also correctly 284 located the critical capacitance where the SNL bifurcations occurs (Fig. 5 H). In the recorded DGGCs, we found no 285 evidence of frequency doubling in the tested capacitance range (Fig. 4 D and F), indicating that the DGGCs are at a 286 point in capacitance space further away from the SNL bifurcation. 287

288 **3 Discussion**

The dynamic clamp is a valuable tool in intracellular recordings to examine the diverse roles of ionic conductances in 289 excitable cells [9, 15, 16, 36]. In this study, we introduced the capacitance clamp (CapClamp), an application of the 290 dynamic clamp that allows to mimic a modified membrane capacitance in a biological neuron. Via simulations of a 291 biophysical neuron model, we confirmed that the CapClamp correctly captures how capacitance affects spike shapes and 292 firing frequency. In recordings of rat dentate gyrus granule cells, we further verified that the CapClamp could accurately 293 control the capacitance of the recorded somatic compartments. Moreover, we clamped this near-somatic capacitance of 294 DGGCs during spiking and found that, as predicted by our simulations, capacitance can modify the fI curve and alter 295 the course of the spike generating currents. CapClamp can serve as a new probe to neuronal signaling, physiology and 296 bifurcations. In the following, we highlight requirements for the CapClamp and discuss how this experimental control 297 over capacitance can benefit the study of cellular electrical behavior. 298

299 3.1 Precise, flexible and local control over capacitance in all excitable cells

To our knowledge, the CapClamp is the first tool to experimentally study capacitance changes in a precise and flexible manner. The CapClamp owes its precision and flexibility to the virtual nature of the altered capacitance. In contrast, methods to physically modify the capacitance are affected by various undesired side effects. Dendritic pinching, decoupling dendrites from the soma, for instance greatly reduces membrane area and thereby capacitance, but also removes all dendritic conductances [37]. Capacitance alterations have also been reported after application of mefloquine, a drug binding to membrane phospholipids, but it also blocks gap junctions [38]. A notable exception is the recent demonstration of engineered polymer synthesis in neuronal cell membranes, which alters their capacitance, but not their input resistance [39]. In comparison, however, the CapClamp provides provides more accurate and dynamic control by allowing to test multiple selected capacitance values in a single cell (Fig. 3 and Fig. 4), while being significantly simpler to implement.

The CapClamp can be applied in every excitable cell. Here, we focused on neurons, but the proposed clamping currents 310 can also be used to study capacitance changes in other cells, including for example heart cells [36, 40]. In particular, 311 no prior knowledge about the ionic or external currents in the clamped cell is required, so that the capacitance can be 312 clamped during any experimental (step current, ramp current, etc.) or during synaptic input. Furthermore, capacitance 313 can be clamped in both electrotonically compact cells like oocytes [41] and non-compact cells like most neurons [26], 314 although in the latter case the CapClamp is limited locally to the capacitance of the recorded compartment (Fig. 3). 315 Consequently, the effects of clamping capacitance depend on the recording site and the cell's morphology. The soma, for 316 example, represents the major capacitive load for spike generation in vertebrate neurons, where the axon predominantly 317 emerges close to the soma (Fig. 4), but it is expected to exert less influence in neurons, where the axon comes out of the 318 dendritic tree, a common feature of invertebrate neurons [42]. 319

The major prerequisite to apply the CapClamp is a reliable capacitance measurement of the clamped compartment, 320 which can be challenging, especially for electrotonically complex cells [1, 2]. An imprecise capacitance estimate leads 321 to erroneous clamping currents, which increase high frequency noise for small errors and might even induce instabilities 322 for larger errors. The measurement method presented for the recorded DGGCs, i.e. mapping the charging response to a 323 two compartment circuit, could in principle be extended to cells with a larger number of compartments e.g. pyramidal 324 cells [43, 44]. Yet, accurate multi-exponential fitting is demanding and the assumption of uniform membrane properties 325 underlying the mapping is a simplification, shown to be violated in some cells, such as GABAergic interneurons 326 [45]. As an alternative, measurement protocols could be exploited that inherently yield local capacitance estimates, 327 including fast voltage ramps [1] or sampling of voltage responses to fast fluctuating currents [46]. Reliable capacitance 328 measurements further allow to compare measured and target capacitance of the clamped cell, which serves as a first 329 simple test to ensure the quality of the CapClamp. 330

331 3.2 A CapClamp on every rig

As a novel application of the established dynamic clamp technique, the CapClamp is an accessible and low-cost extension of a standard electrophysiology stack [15, 16]. For an existing dynamic clamp setup, the sole requirement is to implement the calculation of the clamping currents (see Eq. 5). Otherwise, multiple open source frameworks exist that only require a dedicated computer with a data acquisition card to enable the dynamic clamp in a conventional electrophysiology setup [47–53]. To facilitate the usage of the technique, we provide code for the CapClamp scheme in the RELACS and RTXI frameworks (see Appendix).

In CapClamp recordings, as in all dynamic clamp applications, a high sampling frequency and accurate voltage 338 monitoring are key [54]. Whether a sampling frequency is sufficiently high can be tested by assuring that the observed 339 voltage dynamics e.g. the spike amplitudes are invariant when the sampling frequency is decreased from the maximal 340 possible value [8]. For the simulated fast-spiking interneuron, we found a satisfactory clamp at a frequency of 20 kHz, 341 which we expect to also be sufficient for most excitatory neurons, because they tend to have slower voltage dynamics 342 [20]. In our single electrode recordings, we focused on careful electrode compensation to avoid electrode artifacts in 343 the recorded voltages which would lead to incorrectly estimated membrane currents and eventually instabilities. To 344 improve voltage monitoring, future applications could either apply active electrode compensation [55, 56] or resort to 345 two electrode recordings, where current injection and voltage recordings are separated. 346

347 3.3 Modifying capacitance as a probe for cellular electrical dynamics

Via the CapClamp, experimenters can ask a question that was previously accessible only in theoretical work: What if capacitance was different? In contrast to the theoretical approach, the answers to this question do not have to rely on models of channel dynamics or other membrane properties, because the latter are provided by the biological cell itself [9]. Modifying capacitance with the CapClamp can serve either to investigate changes in this biophysical parameter or, more broadly, to alter the membrane time constant of a cell as a way to characterize its electrical dynamics.

353 3.3.1 Understanding the role of capacitance

The virtual capacitance changes induced by the CapClamp could serve to address two crucial questions about actual 354 membrane biophysics: why capacitance appears to be biologically mostly constant [3] and how exceptions to this rule 355 can facilitate or deter neuronal function [4, 6, 31, 57]. Capacitance is for example rarely tested for optimality - a common 356 question in ion channel kinetics, which appear optimized for function and energy expenditure [20, 58]. Regarding 357 energy consumption, our CapClamp experiments in DGGCs indicate that action potentials become energetically cheaper 358 at lower capacitances (Fig. 5 E and F). Interestingly, reports of exceptional capacitance values mostly find reductions 359 e.g. for myelinated axons ($C_m \approx 0.05$ uF/cm2 for a ten-fold wrapped myelin sheath, see [59]) or human pyramidal cells 360 $(C_m \approx 0.5 \text{ uF/cm}^2, \text{see [6]})$ suggesting that indeed the metabolic cost of AP generation could have been a contributing 361 factor to capacitance adaptations. Moreover, the recent hypothesis that perineuronal nets can reduce capacitance of 362 interneurons in a similar way as myelination of axons suggests that capacitance adaptation could be more widespread in 363 the brain than often assumed [31]. 364

365 3.3.2 Altering the membrane time constant

A key contribution of the CapClamp is the isolated experimental control of the membrane time constant. Monitoring how the membrane potential dynamics change in response to such a perturbation of the time constant has been a theoretical tool to characterize a cell's electrical behavior [11, 13]. As an experimental analogue, the CapClamp could for example be used to further constrain and improve fitting of conductance-based neuron models [11, 60, 61]. Furthermore, the

CapClamp can be used to identify capacitance values where qualitative changes of activity occur (bifurcations, see 370 Fig. 5 G and H), such as a switch to bistable firing [13, 14]. If a neuron is found to be close in capacitance/membrane 37 time constant space to such a critical switch, this can have important implications for infrared [62, 63] and ultrasonic 372 [64, 65] stimulation of neural activity, whose effects are assumed to rely on rapid alteration of the capacitance, as well 373 as for other perturbations such as changes of temperature [13] or extracellular potassium [14], which have similar 374 temporal effects as capacitance. Finally, the broad impact of the time constant on firing frequency and spike shape could 375 be used to examine activity-dependent physiological processes such as ion concentration dynamics [14] or calcium 376 controlled channel homeostasis [66-68]. 377

378 3.4 Conclusion

Taken together, the presented CapClamp enables an accurate and flexible control over capacitance in biological neurons, a basic determinant of cellular excitability, that so far has been inaccessible in experiment. We expect that the CapClamp will, therefore, broaden and enrich the electrophysiological study of neurons and other excitable cells. With expanding techniques to sense and manipulate neural activity, the combination of modeling and targeted closed-loop feedback that underlies the CapClamp (and more generally the dynamic clamp [69]) will further unlock experimental control over other previously inaccessible aspects of single neuron [70–72] and network dynamics [73, 74].

385 4 Materials and Methods

386 4.1 Derivation of the CapClamp current

In order to derive a dynamic clamp feedback scheme for the CapClamp, we compare the actual membrane potential dynamics at the original capacitance C_c with the target dynamics at the chosen capacitance C_t . The actual dynamics of the cell, which for the moment is assumed to be isopotential, is given by the current-balance equation of a single compartment

$$\frac{\mathrm{d}V}{\mathrm{d}t} = \frac{I(V,t) + I_{dyn}(t)}{C_c},\tag{2}$$

with capacitance C_c , membrane currents I(V, t) (comprising all ionic and synaptic currents, as well as external stimuli) and the dynamic clamp current $I_{dyn}(t)$. Note that ionic and synaptic contributions to the membrane currents I(V, t) are voltage-dependent, both with respect to driving force and gating dynamics, so that a voltage trajectory governed by a different capacitance also leads to a modified shape of the membrane currents. In the target dynamics, the dynamic clamp current is absent and the capacitance is modified to the desired value

$$\frac{\mathrm{d}V}{\mathrm{d}t} = \frac{I(V,t)}{C_t}.$$
(3)

Both membrane potential trajectories would coincide, if we chose a dynamic clamp current such that the right-hand sides of actual (Eq. 2) and target dynamics (Eq. 3) become identical,

$$I_{dyn}(t) = \frac{C_c - C_t}{C_t} I(V, t).$$

Generally, an exact model for the membrane currents I(V, t) will not be available, as it would require knowledge about all active conductances and incoming synaptic inputs. Instead, the membrane current can be estimated from the stream of incoming voltage data using the discrete version of Equation 2

$$I(V_{i-1}, t_{i-1}) \approx C_c \frac{V_i - V_{i-1}}{\Delta t} - I_{dyn, i-1}$$
(4)

where Δt is the sampling interval¹. A prerequisite is the measurement of the cell capacitance C_c . Furthermore, for the estimation to be accurate, the sampling interval needs to be shorter than the fastest time scales of changes in the

$$I_{dyn,i} = \frac{C_c - C_t}{C_t} \left(C_c \frac{V_i - V_{i-1}}{\Delta t} - I_{dyn,i-2} \right).$$

¹The indexing in Equation 5 assumes a voltage sampling $V_i = V(i\Delta t)$ and a current injection $I_{dyn,i} = I_{dyn}(i\Delta t)$. However, sampling can take a non-negligible amount of time, so that depending on the sampling system the currently available voltage actually represents the voltage from the previous cycle $V_i = V((i-1)\Delta t)$. In this case, for a correct estimation of the membrane currents, the dynamic clamp current index has to be shifted correspondingly,

membrane currents e.g. sodium gating time constants. With this estimated membrane current, the complete expression
 for the CapClamp current reads

$$I_{dyn,i} = \frac{C_c - C_t}{C_t} \left(C_c \frac{V_i - V_{i-1}}{\Delta t} - I_{dyn,i-1} \right).$$
(5)

The above derivation assumes that the cell is isopotential. In the case of an electrotonically non-compact cell, the steps are identical, but the cell capacitance C_c has to be replaced by the capacitance of the compartment where the recording electrode is located. Consequently, in a non-isopotential neuron, the mimicked capacitance modification is restricted to the compartment at the tip of the recording electrode - a constraint known as the space clamp that is shared by all clamping techniques [15, 75].

396 4.2 Capacitance measurements

To apply the CapClamp, a prerequisite is to measure the capacitance of the recorded local compartment. Here, we use the current clamp protocol, which estimates the capacitance from the voltage response to a current step with amplitude I_{ext} ,

$$V(t) = \sum_{i} v_i \left(1 - e^{-\frac{t}{\tau_i}} \right) = I_{ext} \sum_{i} R_i \left(1 - e^{-\frac{t}{\tau_i}} \right), \tag{6}$$

where an ordering in terms of these time scales is assumed i.e. $\tau_0 > \tau 1 > \dots$ Depending on the morphology, this sum can have a large number of components [76], but in practice often only two or three components can be reliably extracted. As described in Golowasch et al. [1], the slowest component τ_0 is the membrane time constant and allows to infer the total capacitance of a neuron by $C = \frac{\tau_0}{R_0} = \frac{\tau_0}{v_0} I_{ext}$. In the case of an isopotential cell, the membrane time constant is the only component in the charging curve and the total capacitance can be used for the CapClamp.

402 4.2.1 Measurement of near capacitance

For the case of two components τ_0 , R_0 and τ_1 , R_1 in the charging curve (Eq. 6), an equivalent two compartment circuit can be identified comprising a near compartment with capacitance C_n and resistance R_n connected via a coupling resistance R_a to a far compartment with capacitance C_f and resistance R_f [1, Appendix A]. With the additional assumption of a uniform membrane time constant $\tau_m = R_n C_n = R_f C_f$, the fitted two components can be mapped to the values of these five circuit parameters, which in particular provides the near capacitance C_n required for the CapClamp (for the full mapping see [Appendix][Mapping a charging curve with two components to a two compartment circuit])

$$C_n = \frac{\tau_0 \tau_1}{\tau_1 R_0 + \tau_0 R_1}.$$
(7)

When the capacitance is subsequently clamped to a k-fold different value, $C_t = kC_n$, the uniformity assumption has to be correspondingly adjusted to $R_nC_n = kR_fC_f$.

405 4.3 CapClamp in dentate gyrus granule cells

406 4.3.1 Electrophysiology

Acute brain slices were produced as described earlier [77]. Briefly, rats were anesthetized (3% Isoflurane, Abbott, 407 Wiesbaden, Germany) and then decapitated. Brains were removed quickly and transferred to carbogenated (95% O2 / 408 5% CO₂) ice-cold sucrose-ACSF containing (in mM): 87 NaCl, 2.5 KCl, 25 NaHCO₃, 1.25 NaH₂PO₄, 25 glucose, 409 75 sucrose, 7 MgCl₂, 0.5 CaCl₂, 1 Na-pyruvate, 1 ascorbic acid. Horizontal brain slices of 300 µm thickness were cut 410 using a Vibratome (VT1200 S, Leica, Wetzlar, Germany). Hippocampal tissue slices, were collected and placed in 411 a submerged holding chamber filled with carbogenated sucrose ACSF at 32–34 °C for 30 minutes and then at room 412 temperature for 15 minutes before recording. Experiments were alternated between left and right hemisphere slices to 413 prevent bias due to slice condition. 414

For recording, slices were transferred to a submerged chamber and superfused with pre-warmed, carbogenated ACSF containing (in mM): 125 NaCl, 2.5 KCl, 25 NaHCO₃, 1.25 NaH₂PO₄, 25 glucose, 1 MgCl₂, 2 CaCl₂, 1 Na-pyruvate, 1 ascorbic acid. The bath temperature was set to 32–34 °C with a perfusion rate of 12–13 ml/min. Slices were visualized using an upright microscope (AxioScope; Zeiss) equipped with infrared differential inference contrast optics and a digital camera (Retiga EX QImaging CCD, Teledyne Photometrics, AZ, USA). Granule cells from the DG were chosen based on their anatomical location within the hilus as well as their morphological appearance.

Whole-cell patch-clamp electrodes were produced from borosilicate glass capillaries (outer diameter 2 mm, inner 421 diameter 1 mm, Hilgenberg, Germany) using a horizontal puller (P-97, Sutter Instruments, CA, USA) and filled with an 422 intracellular solution consisting of (in mM): K-gluconate 130, KCl 10, HEPES 10, EGTA 10, MgCl₂ 2, Na₂ATP 2, 423 Na₂GTP 0.3, Na₂Creatine 1 and 0.1% biocytin (adjusted to pH 7.3 and 315 mOsm), giving a series resistance of 424 $2.5-4 \,\mathrm{M}\Omega$. All recordings were performed with a SEC LX10 amplifier (npi electronic, Germany), filtered online at 425 20 kHz with the built-in Bessel filter, and digitized at 20 kHz (National Instruments, UK). Following breakthrough into 426 whole-cell configuration, we adjusted the bridge and capacitance compensation before switching to the dynamic clamp 427 mode for recording. Cells were excluded if resting membrane potential was more depolarized than $-45 \,\mathrm{mV}$. The liquid 428 junction potential was not corrected. 429

Neuronal visualization and immunohistochemistry Following recording, selected cells were immersion fixed in 4% paraformaldehyde (PFA) in 0.1 M phosphate buffer (PB, pH 7.4) at 4°C for 24 - 48 hours, slices were then transferred to fresh PB. Prior to immunohistochemical processing, slices were rinsed in PB, followed by PB buffered saline (PBS, 0.9% NaCl). Slices were then rinsed in PBS and incubated in a fluorescent-conjugated streptavidin (Alexa Fluor-647, 1:1000, Invitrogen, UK) in PBS solution containing 3% NGS, 0.1% TritonX-100 and 0.05% NaN3 for 24 hours at 4°C. Slices were rinsed in PBS and then desalted in PB before being mounted (Fluoromount-G, Southern Biotech) on 300 µm thick metal spacers, cover-slipped, sealed, and stored at 4°C prior to imaging.

Confocal imaging and reconstruction DGGCs were imaged on a laser scanning confocal microscope (FV1000, Olympus, Japan). First, a low magnification (4x air immersion, Olympus, Japan) overview image was taken to confirm the cellular type and localization to the DG, then high resolution z-stacks were made with a 30x silicone immersion lens (N.A. 1.05, UPlanSApo, Olympus) over the whole extent of the cell (1 µm axial steps). Images were reconstructed offline using the FIJI software package (imagej.net) and Neutube (neutracing.com)[78]. Image stacks were stitched in FIJI, then the cells were reconstructed and volume filled using Neutube.

Dynamic clamp setup Data acquisition and dynamic clamp loop were controlled by RELACS, V0.9.8, RRID:SCR_017280 using a dedicated computer with a Linux-based real time operating system (rtai.org). The sampling frequency was set to 20 kHz and the recordings were performed in discontinuous current clamp with a duty time of 16.5 μ s. We implemented a CapClamp procedure for RELACS that allows the user to online specify the measured capacitance C_c and the desired target capacitance C_t (for documentation and install instruction see Appendix).

448 **4.3.2** Online measurement of capacitance

For the online measurement of the local capacitance, DGGCs were subjected to twenty hyperpolarizing pulses of 449 $200 \,\mathrm{ms}$ length with $400 \,\mathrm{ms}$ pauses and an amplitude chosen to produce a response of $-5 \,\mathrm{mV}$ in order to minimize 450 interference from active ionic currents. Responses were averaged and the resulting mean trajectory was fit with a sum 451 of exponentials using the Levenberg-Marquardt method from the python library scipy [79]. Fits were performed with 452 one, two and three components and were compared via the F-statistic [80]. In all recorded DGGCs, the two component 453 fit was significantly better than the one exponential fit (p < 0.05, 18/18), whereas no cell exhibited a significant third 454 component (p < 0.05, 0/18). Finally, the extracted two components were mapped to a two compartment circuit as 455 explained above and the near capacitance was then used in the subsequent CapClamp. 456

An offline reexamination revealed that in several recorded cells the above fitting procedure vielded inaccurate estimates 457 of the exponential components, e.g. very short fast components due to an artefactual voltage dip before pulse onset. To 458 circumvent these problems, improved offline fits were performed for the artifact-free recharging at the pulse end (for 459 more details see Appendix). In 8/18 cells, the offline and the original online estimate of the near capacitance differed 460 by less than 20%, but overall the offline measurement yielded higher capacitance values than originally used for the 461 CapClamp (offline: 21.0 ± 9.4 pF, online: 14.9 ± 4.8 pF). In contrast to the online measurement, the offline procedure 462 reported a better fit with three components for a subset of cells (p < 0.05, 7/18), but for the analysis presented here the 463 result of the two component fit is used in all cells. 464

Table 2: Multi-exponential fit and corresponding circuit parameters in the recorded dentate gyrus granule cells (N=18) and a multicompartment model based on a reconstructed DGGC morphology (see Simulations).

	DGGCs (mean \pm std)	multicomp. model
Exp. fit		
$ au_0$	$15.1\pm4.8~\mathrm{ms}$	15.1 ms

	DGGCs (mean \pm std)	multicomp. model
R_0	$127.1\pm44.6~\mathrm{M}\Omega$	119.2 MΩ
$ au_1$	$0.77\pm0.24~\text{ms}$	0.18 ms
R_1	$34.5\pm14.7~\text{M}\Omega$	12.3 MΩ
Circuit		
C_n	$21.0\pm9.4~\text{pF}$	13.0 pF
R_n	$854.2\pm394.0~\text{M}\Omega$	1158.0 M Ω
R_a	$52.5\pm19.8~\text{M}\Omega$	15.5 MΩ
C_f	$105.8\pm33.0\ \text{pF}$	113.7 pF
R_f	$155.5\pm59.9~\text{M}\Omega$	132.8 MΩ

465 **4.3.3** Protocol 1: Verification of altered capacitance

After online measurement of the capacitance, each DGGC was clamped at a range of capacitances from 60% to 300% of the original near capacitance. For each clamped capacitance, the above offline capacitance measurement protocol was repeated to see how the CapClamp altered the slow and fast components. These time scale and amplitude changes were then mapped to the corresponding two compartment circuit parameters to compare them to the target capacitance (see Measurement of near capacitance). Due to the difference between online and offline estimate of the original near capacitance, we corrected the original target capacitance to $C_t^{\text{corr}} = C_c^{\text{off}} + \Delta C_t$, which preserves the targeted capacitance change $\Delta C_t = C_t - C_c^{\text{on}}$. Equally, the clamping factors in the mapping were updated to $k = \frac{C_t^{\text{corr}}}{C_{\text{off}}}$.

473 4.3.4 Protocol 2: Analysis of fI curves and spike shapes

In a subset of cells, after measuring near capacitance, an fI curve was obtained for the original capacitance and for target capacitances in the above range. Current pulses were 1 s long and repeated three times, at amplitudes ranging from 90% to 200% of an estimated rheobase. This rheobase was estimated by the first occurrence of spiking in response to a ramp (length: 5 s, height: 250 pA). For a quantitative comparison, the resulting fI curves were fit by a square-root function

$$f(I) = \Theta(I - I_{rheo}) \operatorname{gain} \sqrt{I - I_{rheo}}$$

474

which captured their type 1 firing with a continuous frequency-current relationship [34].

Spikes were detected as a minimum 10 mV elevation over the average depolarization during the pulse. For the mean action potential (AP) shape, varying spike forms from the initial (< 300 ms) part of the pulse were discarded. The extracted AP features were peak amplitude, threshold voltage and threshold time to peak (voltage derivative crossing 10 mV/ms), height (difference between peak and threshold), temporal width at half of the height and fast

afterhyperpolarization (fAHP; a voltage dip of -0.5 mV or larger within 10 ms after the spike). For threshold and

fAHP detection, the spike shape was filtered with a digital 4th order Butterworth filter with critical frequencies 3.3 kHz,
 respectively 1 kHz.

To understand, how much changes in capacitance affect the action potential generating currents, we compared the recorded spikes with hypothetical ones obtained by assuming unaltered currents with respect to the original near capacitance. For a target capacitance C_t , this hypothetical spike is a scaled version of the original spike,

$$V_{hypo}(t) = V_c(t_0) + \frac{C_c}{C_t} \left(V_c(t) - V_c(t_0) \right)$$

where $V_c(t)$ is the spike form at the original cell capacitance C_c and the initial time t_0 was chosen to be $t_{spike} - 3ms$ short before onset of the spike generating currents.

485 4.4 Simulations of the CapClamp

Simulations of neuron models coupled to the CapClamp were implemented using the neuron simulator Brian2 [81]
and the CapClamp was realized using the Brian2 provided NetworkOperation that updated the clamp current every
sampling interval using Equation 5 with zero delay between voltage sampling and current injection (for links to the
available code see Appendix).

490 4.4.1 Biophysical neuron model

In order to test the CapClamp in the presence of active ionic conductances, a Wang-Buzsáki (WB) neuron, a single 491 compartment model of hippocampal interneurons, was used [23]. Gating dynamics and peak conductances of the 492 transient sodium current and the delayed rectifier potassium current were modeled as described earlier [13, Appendix 493 A]. When the capacitance is varied, the WB neuron undergoes a well-characterized series of bifurcations; in particular it 494 exhibits a saddle-node loop (SNL) bifurcation at $C_m = 1.47 \frac{\mu F}{cm^2}$ [13, Fig. 6]. Here, the specific membrane capacitance 495 was chosen as $C_m = 0.75 \frac{\mathrm{uF}}{\mathrm{cm}^2} \frac{\mu \mathrm{F}}{\mathrm{cm}^2}$ and the membrane area was set to $A = 20000 \ \mu \mathrm{m}^2$, so that the original cell 496 capacitance was 150 pF. Simulations were performed with the second order Runge-Kutta method, a time step of 1 μ s 497 and dynamic clamp loop frequencies up to 100 kHz. Analysis of spike shapes was performed in the same way as for the 498 recorded cells. 499

500 4.4.2 Multicompartment model of a dentate gyrus granule cell

For a controlled test of the CapClamp in an electrotonically non-compact cell, a morphologically reconstruction of a recorded DGGC was used as the basis for a multicompartment simulation. Soma and the two dendritic trees had a total area of 14126 μ m². The axon was removed for the simulation. Membrane properties were assumed to be uniform and chosen such that they reproduced the average values of the total capacitance and the membrane time constant observed in the experiments: $C_m = \frac{C_n + C_f}{A} \approx 0.9 \frac{\mu F}{cm^2}$ and $R_m = \frac{\tau_0}{C_m} \approx 16800 \ \Omega cm^2$. The axial resistivity was chosen as $R_a = 300 \ \Omega cm$. Simulations were performed with exponential Euler integration, a time step of 10 μ s and a dynamic

⁵⁰⁷ clamp sampling frequency of 20 kHz. Capacitance measurement and clamp procedure were the same as in the recorded

508 DGGCs.

509 5 Acknowlegdements

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516 6 Competing interests

⁵¹⁷ The authors declare that no competing interests exist.

518 7 Appendix

519 7.1 Impedance of a capacitance-clamped RC circuit

The impedance of a cell captures its response to the whole range of input frequencies (see Fig. 1 - suppl. 1). In the following, we derive the impedance of a passive membrane, an RC circuit, with capacitance C_c coupled to the CapClamp and compare it to the impedance of an RC circuit with the target capacitance C_t .

523 7.1.1 Analysis of the dynamic clamp via the Z-transform

In general, the dynamic clamp technique forms a digital filter, mapping the incoming sampled voltages to injected currents. For a sampling interval Δt , a linear mapping such as the CapClamp has the form

$$I_{dyn}(i\Delta t) = \sum_{j=0}^{N} \nu_j V\left((i-j)\Delta t\right) + \sum_{k=1}^{M} \gamma_k I_{dyn}\left((i-k)\Delta t\right),\tag{8}$$

where N and M determine history of voltage and current values, respectively, taken into account. For the CapClamp, the coefficients depend on cell capacitance C_c , target capacitance C_t and the sampling interval (see Eq. 5),

$$\nu_0 = \frac{C_c - C_t}{C_t} \frac{C_c}{\Delta t},$$

$$\nu_1 = -\nu_0,$$

$$\gamma_1 = -\frac{C_c - C_t}{C_t}.$$
(9)

This linear mapping can be represented and analyzed using the Z-transform [82, Ch. 13],

$$\hat{I}(z) = F_{dyn}(z)\hat{V}(z),\tag{10}$$

where the transfer function follows from the properties of the Z-transform²

$$F_{dyn}(z) = \frac{\sum_{j=0}^{N} \nu_j z^{-j}}{1 - \sum_{k=1}^{M} \gamma_k z^{-k}}.$$
(11)

If the cell also forms a linear system, like the RC circuit, the transfer function of the coupled system (Fig. 6 A) is given by [82, Table 2.6]

$$H_{\text{cell+dyn}}(z) = \frac{H_{\text{cell}}(z)}{1 - H_{\text{cell}}(z)F_{\text{dyn}}(z)},\tag{12}$$

where $H_{\text{cell}}(z)$ is the Z-transform of the membrane filter, e.g. $H_{\text{cell}}(z) = H_{\text{RC}}(z)$.

²A Z-transform $X_i \xrightarrow{Z} \hat{X}(z)$ is linear and has the delay property $X_{i-1} \xrightarrow{Z} z^{-1} \hat{X}(z)$ [82, Table 13.2].

The transfer function of the coupled system $H_{\text{cell}+\text{dyn}}(z)$ can then be compared with the one of the target system $H_{\text{target}}(z)$ (Fig. 6 B). Additionally, the frequency-dependent impedance can be retrieved from the transfer function by

$$Z_{\text{cell}+\text{dyn}}(f) = H_{\text{cell}+\text{dyn}}(e^{i2\pi f\Delta t}).$$
(13)

527 7.1.2 Transfer function of the CapClamp

The Z-transform of the CapClamp filter can be read directly from the general form of the transfer function (Eq. 11) and the CapClamp feedback coefficients (Eq. 9),

$$F_{\rm dyn}(z) = \frac{C_c - C_t}{C_t} \frac{C_c}{\Delta t} \frac{1 - z^{-1}}{1 + \frac{C_c - C_t}{C_t} z^{-1}}.$$
(14)

528 7.1.3 Transfer function of the RC circuit

529 In an RC circuit, the dynamics of the voltage are

$$C\frac{\mathrm{d}V}{\mathrm{d}t} = -\frac{V}{R} + I.$$

530 Thus, in a single time step Δt , when the current is fixed, the voltage evolves as

$$V(k\Delta t) = V((k-1)\Delta t)e^{-\frac{\Delta t}{\tau}} + RI(1 - e^{-\frac{\Delta t}{\tau}}),$$

where $\tau = RC$ is the time constant. Applying the Z-transform results in the transfer function

$$H_{RC}(z) = R\left(1 - e^{-\frac{\Delta t}{\tau}}\right) \frac{1}{z - e^{-\frac{\Delta t}{\tau}}},\tag{15}$$

which is subsequently used as the cell's transfer function $H_{cell}(z) = H_{RC}(z)$.

533 7.1.4 Transfer function of the clamped RC circuit

Introducing $K = \frac{C_c - C_t}{C_t}$ and $h_c = \frac{\Delta t}{\tau_c}$, the RC circuit (Eq. 15) and CapClamp (Eq. 14) transfer functions can be combined using Equation 12 to get the transfer function of the combined system

$$H_{\text{cell+dyn}}(z) = R(1 - e^{-h_c}) \frac{z + K}{z^2 + (K - e^{-h_c} - \frac{1}{h_c}K(1 - e^{-h_c}))z - K(e^{-h_c} - \frac{1}{h_c}(1 - e^{-h_c}))}.$$
 (16)

536 In comparison, the transfer function of the target RC circuit reads

$$H_{\text{target}}(z) = R \left(1 - e^{-h_t}\right) \frac{1}{z - e^{-h_t}}$$

with $h_t = \frac{\Delta t}{\tau_t} = \frac{\Delta t}{RC_t}$ reflecting the different target capacitance.

538 Figure 1 - suppl. 1 compares the resulting impedances for decreased and increased capacitances. As discussed in the

Results, the impedance amplitudes fit well up to a tenth of the dynamic clamp frequency. A closer look at the transfer
 function explains the fit at low frequencies and the deviations at higher frequencies.

Input resistance is preserved The input resistance is equal to the impedance at zero frequency, that is at $z = e^{i2\pi 0} = 1$, which for both coupled and target system is the original resistance,

$$H_{\text{cell}+\text{dyn}}(1) = H_{\text{target}}(1) = R.$$
(17)

Poles and zeros For a further comparison, poles and zeros of the transfer functions are calculated. To simplify the expressions, it is assumed that the time constant of the original and target circuits are much larger than the sampling interval, that is $h_c \ll 1$ and $h_t \ll 1$.

Target circuit The target circuit has no zero and a single pole located at

$$p_t^{(1)} = e^{-h_t} = 1 - h_t + \dots$$
(18)

Capacitance clamped circuit The coupled system has one zero at

$$r_c^{(1)} = -K = 1 - \frac{C_c}{C_t}.$$
(19)

The clamped circuit has two poles at

$$p_c^{(1)} = 1 - (1+K)h_c + \dots$$
(20)

and

$$p_c^{(2)} = \frac{K}{2}h_c + \dots$$
 (21)

Comparison of poles All poles and zeros for an RC circuit in its original state and clamped at decreased and increased capacitances are shown in Figure 6. The first pole of the clamped circuit coincides with the one of the target circuit: $p_c^{(1)} = 1 - \frac{C_c}{C_c} \frac{\Delta t}{RC_c} = 1 - h_t = p_t^{(1)}$. As these pole lies close to z = 1, they determine the lower frequency response, which explains why the impedance amplitudes fit so well in this range.

In addition to moving the existent pole of the cell circuit to the one of the target circuit, the CapClamp creates an additional pole $p_c^{(2)} \approx \frac{h_c}{2} (\frac{C_c}{C_t} - 1)$ and a new zero $r_c^{(1)} = 1 - \frac{C_c}{C_t}$. Thus, at an increased capacitance $C_t > C_c$, the new pole lies in the left half of the unit circle and thereby increases the impedance at higher frequencies. In contrast, at a decreased capacitance, the additional zero moves into the left half of the complex plane and thereby decreases the impedance at higher frequencies.

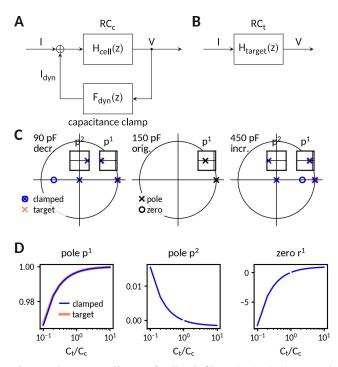


Figure 6: Analysis of the capacitance clamp as a discrete feedback filter. A Block diagram of the coupled system: RC circuit with original capacitance C_c and capacitance clamp feedback current. B Block diagram of the target system: RC circuit with target capacitance C_t . C Pole-zero plot of the transfer functions at a decreased (left), the original and an increased capacitance. In addition to mimicking the pole of the target system, the clamped system has an additional pole and an additional zero. D Pole and zero position versus capacitance.

Stability For the investigated RC circuit with $R=100 \text{ M}\Omega$ and C = 150 pF and a sampling interval of 50 us, both poles of the capacitance clamped system remain within the unit circle (Fig. 6) for the tested range from 0.1 to 10 times the original capacitance. As the coupled system is naturally causal, this implies that the transfer function of the clamped circuit is stable for this range of target capacitances, i.e. there are no unstable oscillations.

557 7.2 Mapping between a charging curve with two components and a two compartment circuit

In the following, we explain how a charging curve of a cell with two components can be mapped to the parameters of a two compartment circuit, which we used to extract the local capacitance in the recorded dentate gyrus granule cells (see Fig. 3). We first report the approach and results derived earlier [1, Appendix] and then explain how to extend the mapping when the capacitance is clamped to a modified value.

Golowasch et al. derived expressions for the near capacitance and the other circuit parameters by comparing the impedance of a two compartment circuit in Figure 3 A

$$Z(s) = \frac{1}{\frac{1}{R_n} + sC_n + \frac{1}{R_a + \frac{1}{\frac{1}{R_f} + sC_f}}}$$
(22)

with the impedance of a system whose response to a step currents is a sum of two exponentials

$$Z(s) = R_0 \frac{1}{1 + s\tau_0} + R_1 \frac{1}{1 + s\tau_1}.$$
(23)

The comparison of these two impedances gives four equations linking the circuit parameters and the two components of the charging curve:

$$R_0 + R_1 = \frac{R_a R_n + R_f R_n}{R_a + R_f + R_n},$$
(24)

$$R_0\tau_0 + R_1\tau_1 = \frac{R_a R_n R_f C_f}{R_a + R_f + R_n},$$
(25)

$$\tau_0 + \tau_1 = \frac{(R_a + R_n)C_f R_f + (R_a + R_f)C_n R_n}{R_a + R_f + R_n},$$
(26)

$$\tau_0 \tau_1 = \frac{R_a C_n R_n C_f R_f}{R_a + R_f + R_n}.$$
(27)

To solve this set of equations, they assume that the membrane time constant is the same in all compartments $C_n R_n = C_f R_f = \tau_c$. However in a clamped neuron, where the near capacitance is targeted to be modified to a k-fold different value, this equation becomes

$$C_n R_n = k C_f R_f, (28)$$

567 where $k = \frac{C_{n,clam.}}{C_{n,orig.}}$.

For the unclamped case, k = 1, the mapping from the two components to the circuit parameters is

$$R_n = R_0 + \frac{\tau_0}{\tau_1} R_1,$$
(29)

$$C_n = \frac{\tau_0 \tau_1}{\tau_1 R_0 + \tau_0 R_1},\tag{30}$$

$$R_f = \frac{R_0 \tau_1}{R_1 \tau_0} \left(R_0 + \frac{\tau_0}{\tau_1} R_1 \right) \tag{31}$$

$$C_f = \frac{R_1 \tau_0}{R_0 \tau_1} \frac{\tau_0 \tau_1}{\tau_1 R_0 + \tau_0 R_1}$$
(32)

$$R_{a} = \frac{\tau_{1}}{\tau_{0} - \tau_{1}} \left(R_{0} + \frac{\tau_{0}}{\tau_{1}} R_{1} \right) \left(1 + \frac{R_{0} \tau_{1}}{R_{1} \tau_{0}} \right).$$
(33)

For the clamped case, $k \neq 1$, we used the python package sympy to solve the equations.

570 7.3 Adapted fitting procedure of dentate gyrus charging curves

⁵⁷¹ The initial online capacitance measurement was based on fitting the charging curve at the beginning of the current pulse.

⁵⁷² Posterior analysis showed an artefactual voltage drop of -0.2 mV starting about 0.2 ms before pulse onset (probably due

to coupling of the DAQ measurement card and the motherboard of the dynamic clamp computer), which limited the reliability of the online fit for cells with a small fast component. As no such artifact was observed for the recharging at the end of the pulse, this part was used in an improved offline fit. Additional measures to improve the fit were: cut of the first 0.2 ms after pulse end to minimize electrode artifacts, limiting the fit to the first 60 ms (3-4 times τ_0) after the pulse to prioritize the early part of the charging curve and a switch to the python package lmfit for better evaluation of parameter confidence bounds (lmfit.github.io). Furthermore, the finite rise time of the current injection by the amplifier was taken into account by adapting the original form of the charging curve (Eq. 6) to

$$V(t) = I_{ext} \left[\sum_{i; \tau_i \neq \tau_a} \frac{R_i}{\tau_i - \tau_a} \left(\tau_i \left(1 - e^{-\frac{t}{\tau_i}} \right) - \tau_a \left(1 - e^{-\frac{t}{\tau_a}} \right) \right) + \sum_{i; \tau_i = \tau_a} R_i \left(1 - e^{-\frac{t}{\tau_i}} - \frac{t}{\tau_i} e^{-\frac{t}{\tau_i}} \right) \right],$$
(34)

where the current rise time of the amplifier τ_a (87 ± 2 µs) was obtained by fitting the recorded injected current for the current step command by a simple exponential. A comparison of the two exponential components and the resulting circuit parameters for the online and offline fitting procedures is shown in Tab. 3.

For the charging curves under capacitance clamp, the fitting procedure for the charging curve with two exponentials was initialized with values as expected for the targeted capacitance change: mapping the fitting results of the unclamped response to a two compartment circuit, changing the near capacitance to the targeted value and finally mapping this altered circuit back to the expected time scale and amplitudes. This initialization improved the fits especially at increased near capacitances, where the amplitude of the fast component becomes smaller.

Table 3: Comparison of online and offline fits to charging curves in the recorded dentate gyrus granule cells (N=18).

	online fit (mean \pm std)	offline fit (mean \pm std)
Two comp.		
$ au_0$	$14.9\pm4.8~\mathrm{ms}$	$15.1\pm4.8~\mathrm{ms}$
R_0	$136.9\pm47.5~\text{M}\Omega$	$127.1\pm44.6~\mathrm{M}\Omega$
$ au_1$	$0.41\pm0.23~\text{ms}$	$0.77\pm0.24~\text{ms}$
R_1	$25.1\pm14.1~\text{M}\Omega$	$34.5\pm14.7~\mathrm{M}\Omega$
Circuit		
C_n	$14.9\pm4.7~\mathrm{pF}$	$21.0\pm9.4~\text{pF}$
R_n	$1106.3\pm519.3~\mathrm{M}\Omega$	$854.2\pm394.0~\text{M}\Omega$
R_a	$34.9\pm19.9~\text{M}\Omega$	$52.5\pm19.8~\text{M}\Omega$
C_f	$99.1\pm33.7~\text{pF}$	$105.8\pm33.0~\text{pF}$

	online fit (mean \pm std)	offline fit (mean \pm std)
$\overline{R_f}$	$159.6\pm58.1~\text{M}\Omega$	$155.5\pm59.9~\text{M}\Omega$

588 7.4 Data and software availability

589	• Electrophysiological recordings of capacitance clamped dentate gyrus granule cells: Paul Pfeiffer, & Federico
590	José Barreda Tomás. (2021). Capacitance clamp demonstration in rat dentate gyrus granule cells (1.0.0) [Data
591	set]. Zenodo. 10.5281/zenodo.5552207
592	• Project repository with capacitance clamp module for RELACS and custom analysis/simulation in python:
593	Paul Pfeiffer, Federico José Barreda Tomás, Jiameng Wu, Jan-Hendrik Schleimer, Imre Vida, & Susanne
594	Schreiber. (2021). Softfware for: A dynamic clamp protocol to artificially modify cell capacitance (v1.0).
595	Zenodo. 10.5281/zenodo.5762092
596	• Capacitance clamp plugin for RTXI, a real-time data-acquistion and control application for biological re-
597	search that allows to extend a conventional electrophysiology setup for dynamic clamp experiments [51]:
598	capacitance_clamp_rtxi_module: (v1.0.1). Zenodo. 10.5281/zenodo.5553946

- 599 7.5 Supplementary figures
- Figure 1 suppl. 1: Impedance analysis of an RC circuit coupled to the capacitance clamp.

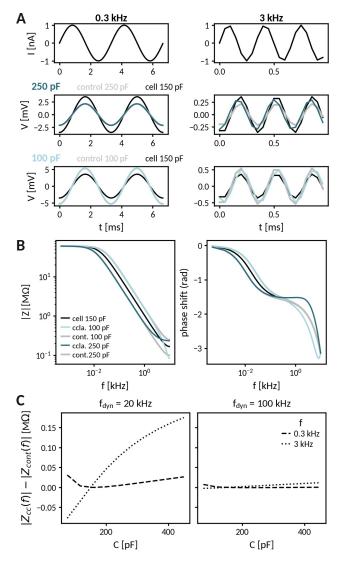


Figure 1 - suppl. 1: **Impedance analysis of an RC circuit coupled to the capacitance clamp.** A Injection of an oscillating current at 300 Hz (left) and at 3 kHz (right) to a passive cell (RC-circuit) with voltage responses clamped at an increased (middle) and a decreased capacitance (bottom). Black lines indicate the response of the cell at the original capacitance and gray lines those of the corresponding control cells. B Comparison of frequency-dependent impedance and phase shift of a cell at the above capacitances (black: cell capacitance, blues: clamped, gray: control). C Difference of impedances at 300 Hz (dotted) and 3 kHz (dashed) for clamped and control cell across different capacitances and for dynamic clamp frequencies of 20 kHz (left) and 100 kHz (right).

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